

# LPR30

by T. Sampieri Standard Circuit Product Marketing

## LOW DROP VOLTAGE REGULATOR DRIVER FOR AN EXTERNAL MOSFET

The **LPR30** is a very Low Dropout Regulator Controller in a single IC solution for very high current, low dropout linear voltage regulation. It uses an external N-channel Power MOSFET as the linear pass element. The **LPR30** features a dropout voltage as low as the  $R_{DS(on)}$  of the external Power MOSFET multiplied by the output current. Consequently, the output current can be as high as the Power MOSFET can provide (using an adequate heatsink).

The  $V_{CC}$  of the **LPR30** ranges from 5V to 30V. For very low dropout voltage operation, the **LPR30** requires an external gate drive supply to provide the control voltage needed to drive the gate of the external Power MOSFET.



The regulator output is constant-current limited when the controller detects 50mV across an external sense resistor (R<sub>SENSE</sub>). It has an internal high precision

> (+/- 1%) Voltage Reference at 2.5V. The output regulated voltage is programmable from 1V to 50V. Flexible design is achieved by the availability of the Voltage Reference Output through an external pin (n.4) that is able to supply more than 20mA as a load current. The LPR30 is available in 8-pin plastic DIP and in SO-8 for SMD. In both package versions it is able to operate from 0°C to 70°C.



Typical application





### FEATURES

- Very low dropout Power MOSFET driver
- High precision  $V_{ref} = 2.5V (+/-1\%)$
- Very low current drain (Typ. 2mA)
- Reference output current > 20mA
- Operating supply voltage from 5V to 30V
- Max V<sub>in</sub> on gate (pin n.8) up to 60V
- Internal current limit operational amplifier
- offset trimmed at 50mV + -4mV
- No capacitor is needed for stability of reference output
- Temperature range: 0°C to 70°C

Туре	Package
LPR30N	DIP8
LPR30D	SO-8
LPR30D-TR	SO-8 IN TAPE & REEL

### **APPLICATIONS**

Ultra-high current, ultra-low dropout voltage regulator

- Constant high current source
- Low part count 5.0V to 3.3V computer supply
- Low noise/low drop SMPS post regulator
- Motherboard



# LPR30 DEMOBOARD

The schematic that follows is of a fixed voltage regulator demoboard developed to show the optimum performance of the LPR30.



LPR30 demoboard circuit

If a  $V_{out}$  >25V is required, resistors R8 and R11 and capacitor C5 can be omitted and pin 4 and 6 of the IC are connected together.

![](_page_1_Picture_26.jpeg)

![](_page_2_Picture_0.jpeg)

The capacitor C5 can be left to reduce noise. In order to allow the complete Power MOSFET driving, V2 must be greater than  $V_{out} + V_{(GS)sat}$ .

If V1 is predicted to be greater than  $V_{out}$  +  $V_{(GS)sat}$  it is possible to use a single supply, connecting V1 and V2 together. The use of a logic level Power MOSFET (such as STB80N03L-06) lowers the V2 voltage (V<sub>in</sub> must be less than 30V).

If the  $V_{out}$  is forseen to be less than 30V, R9 and R10 can be omitted and pin 2 of the IC can be connected directly at the R7 resistor (Vout side).

The waveform below shows the settling time with a load variation from 0mA to 1500mA. It is evident that the response is limited by R2, R3 and C4.

R3-C4 net is chosen to give a compromise between response speed and stability. C3 also improves the system stability.

The waveform on the right shows the maximum current that the voltage reference is able to supply (pin n. 4).

A datailed description of this demoboard is available in an Application Note.

![](_page_2_Figure_9.jpeg)

Load variation from 0mA to 1500mA

![](_page_2_Figure_11.jpeg)

Solder side of demoboard

![](_page_2_Figure_13.jpeg)

Component side of demoboard

![](_page_2_Figure_15.jpeg)

![](_page_2_Figure_16.jpeg)

![](_page_2_Picture_17.jpeg)

![](_page_3_Picture_0.jpeg)

# LPR30

by T. Sampieri - Standard Circuits Product Marketing & F. Lentini - Application Engineer

### LOW DROP VOLTAGE REGULATOR FOR AN EXTERNAL POWER MOSFET NEW APPLICATIONS

n the previous edition of the Express we introduced the new extremely fast low drop voltage regulator driver for an external (N-channel) Power MOSFET. As part of our on going promotion we have developed new applications for this innovative driver.

#### INCREASED RESPONSE SPEED

The response of the LPR30 is already very fast, but with some very small changes it is possible to greatly improve its performance. In the standard configuration the total response is limited by the compensation network (R2, R3 and C4), which limits the voltage rise-time at the gate of the external Power MOSFET.

Improved stability performances are given with R3=0 $\Omega$  and C4=2.2 $\mu$ F electrolytic capacitor. With the preset value of C4, the response time is determinated by R2.

![](_page_3_Figure_9.jpeg)

This value must be chosen for a current no higher than 100mA on pin 8 of the LPR30. For instance with V2=12V, we suggest a R2 value of 100 $\Omega$ . In this case, under static conditions:

 $I_{R2} = VR2/R2 \cong (V2 - V_{GS} - V_{out})/R2$   $I_{R2} \text{ reaches the maximum value in short}$ circuit conditions when V<sub>out</sub>=0V and V<sub>GS</sub> is minimum (near the threshold value).

![](_page_3_Figure_12.jpeg)

![](_page_3_Picture_13.jpeg)

![](_page_4_Picture_0.jpeg)

With these values and V<sub>out</sub>=5V, I<sub>R2</sub> is 20mA in normal conditions and becomes 100mA under short circuit conditions.

In this case maximum Power dissipated on R2 is :

 $P_{R2} = R2 * (I_{R2max})^2 = 1W$ If V2 is higher than 12V, the power dissipated on R2 could be too high.

### A VERSATILE SOLUTION

The use of a small NPN transistor (such as BC337) could solve the problem of a high dissipation value.

This transistor would work only during the transition of the load, increasing the total reaction speed, but in static situations it will be "off", opening the connection to ground, so the resistor R2 will not dissipate any power.

For this reason it is possible to reduce the value of the resistor to  $50\Omega$  further improving the total response speed without sacrificing any power.

### FASTER THAN COMPETITION

A leading manufacturer of low drop Voltage Regulators publicized the optimum speed of its LT1585 (output current 4.6A) Under identical conditions our LPR30 is roughly 50% faster !

![](_page_4_Figure_11.jpeg)

Туре	Package
LPR30N	DIP8
LPR30D	SO-8
LPR30D-TR	SO-8 IN TAPE & REEL

![](_page_4_Figure_13.jpeg)

![](_page_4_Figure_14.jpeg)

![](_page_4_Figure_15.jpeg)

R2= 50  $\Omega$  Cout= 220  $\mu\text{F}$ 

![](_page_4_Figure_17.jpeg)

![](_page_4_Picture_18.jpeg)

![](_page_5_Picture_0.jpeg)

The following waveforms show the different behaviour of the demoboard configured to supply a Vout=5V with a maximum supply current up to Iout= 7A. It can clearly be seen that the transient time required to recover the Vout when we put a high load is always very short, from 15µs (for I<sub>out</sub> from 0 to 4A) to maximum 60µs (for I<sub>out</sub> from 0 to 7A).

![](_page_5_Figure_3.jpeg)

### LOAD VARIATION

The load variation is also dependent on the Cout capacitance value (C3). With a standard electrolytic capacitor of 1000µF, the maximum load variation is around 65mV and under the same conditions, with a maximum load of 7A and a smaller capacitor of 220µF, the load variation is still very low (150mV).

No other competitor is able to supply a device with such excellent performance!

## LPR30 DEMOBOARD

A promotional kit containing a demoboard and floppy disk is available on request.

![](_page_5_Picture_9.jpeg)

![](_page_5_Picture_10.jpeg)